

## Lecture 2 (Chapter 2) Discrete-Time Signal and Systems

### Classification of Signals

1. Finite duration  $x(n) = 0 \quad n > N$

Infinite duration

2. Right-Left sided

$x(n) = 0 \quad n < N_1$  right-sided

$x(n) = 0 \quad n > N_2$  left-sided

Some Elementary Discrete-Time Signals

- unit sample sequence  $\sigma(n) = \begin{cases} 1 & n = 0 \\ 0 & n \neq 0 \end{cases}$

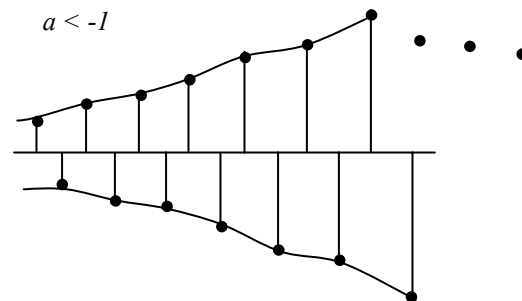
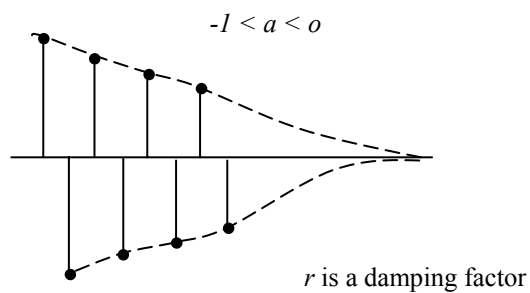
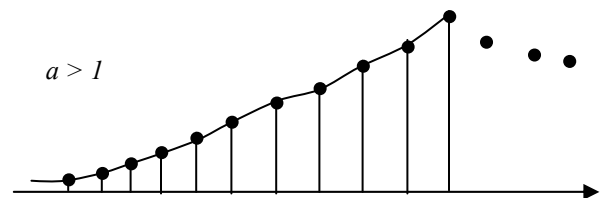
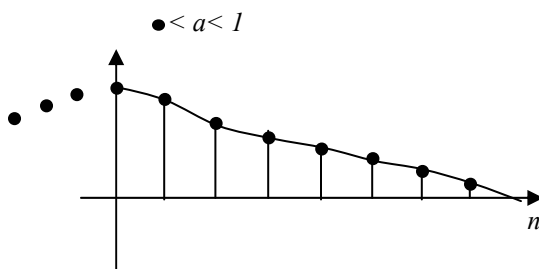
- unit step sequence  $u(n) = \begin{cases} 1 & n \geq 0 \\ 0 & n < 0 \end{cases}$

- unit ramp  $u_r(n) = \begin{cases} n & n \geq 0 \\ 0 & n < 0 \end{cases}$

- exponential  $x(n) = a^n$  for all  $n$

If  $a$  is complex, then  $a = re^{j\theta} \rightarrow x(n) = r^n e^{j\theta n}$

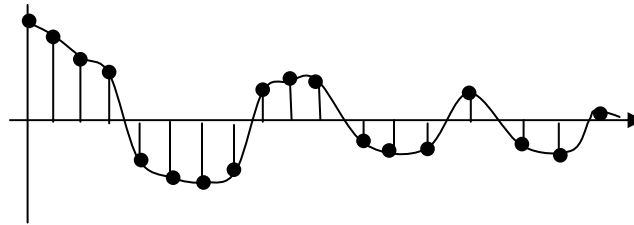
$$= r^n (\cos \theta n + j \sin \theta n)$$



For complex  $a$ ,  $x(n) = x_R(n) + jx_I(n)$ , where  $x_R(n) = r^n \cos(\theta n)$

if  $r < 1$  then

$$\begin{cases} |x(n)| = r^n \\ \angle x(n) = \theta n \end{cases}$$



### Energy and Power Signals

Energy is defined as  $E = \sum_{-\infty}^{+\infty} |x(n)|^2$  if  $E$  is finite, i.e.,  $0 < E < \infty$ , then  $x(n)$  is called

Energy Signal. However, many signals that have an infinite energy, have a finite average power. Average power is defined as

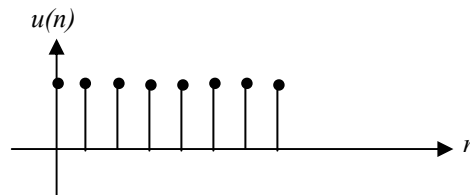
$$P_{ave} = \lim_{N \rightarrow \infty} \frac{1}{2N+1} \sum_{-N}^N |x(n)|^2.$$

If we define the signal energy of  $x(n)$  over the interval  $(-N, N)$  as

$$E_N = \sum_{-N}^N |x(n)|^2 \quad \text{then} \quad E = \lim_{N \rightarrow \infty} E_N$$

and therefore,  $P_{ave} = \lim_{N \rightarrow \infty} \frac{1}{2N+1} E_N$  clearly if  $E$  is finite, then  $P_{ave} = 0$ .

### Example – Unit Step Sequence



Obviously, it is not an energy signal but it is a power signal.

$$\begin{aligned}
 p &= \lim_{N \rightarrow \infty} \frac{1}{2N+1} \sum_{-N}^N |x(n)|^2 = \lim_{N \rightarrow \infty} \frac{1}{2N+1} \sum_{n=0}^N (1)^2 \\
 &= \lim_{N \rightarrow \infty} \frac{N+1}{2N+1} = \frac{1}{2} \Rightarrow \text{it is a power signal!}
 \end{aligned}$$

### **Periodic and Aperiodic Signals**

$$x(n+N) = x(n) \Rightarrow \text{signal is periodic}$$

Energy of periodic signals is infinite but it might be finite over a period. On the other hand, the average power at the periodic signal is finite and is equal to the  $P_{ave}$  over a single period. Hence, periodic signals are power signals.

### **Symmetric (even) and odd Signals**

$$x(n) = x(-n) \text{ even}$$

$$x(n) = -x(-n) \text{ odd}$$

Any signal can be written as:  $x(n) = x_e(n) + x_o(n)$

$$\text{Where } \begin{cases} x_e(n) = \frac{1}{2} [x(n) + x(-n)] \\ x_o(n) = \frac{1}{2} [x(n) - x(-n)] \end{cases}$$

Read Section 2.1.8

## **Classification of Discrete-Time Systems**

### **Static versus Dynamic Systems**

*Static Systems*  $\equiv$  memory less  $\equiv$  the output doesn't depend on past or future values of the input.

*Dynamic Systems*  $\equiv$  having either infinite or finite memory.

Example:  $y(n) = 2x(n) + x(n)^2$  Static

$$y(n) = \sum_{k=0}^N x(n-k) \text{ Dynamic-finite}$$

$$y(n) = \sum_{k=0}^{\infty} x(n-k) \text{ Dynamic-infinite}$$

## Time invariant versus Time-Invariant Systems

A relaxed system  $\Gamma$  is time-invariant if

$$x(n) \rightarrow y(n) \\ \forall K, x(n-k) \rightarrow y(n-k)$$

Example: 1)  $y(n) = x(n) - x(n-1)$

$$y(n-k) = x(n-k) - x(n-k-1) \quad \text{time invariant}$$

2)  $y(n) = nx(n)$

$$y(n-k) = (n-k)x(n-k) = nx(n-k) - kx(n-k)$$

$$\text{but } x(n-k) \rightarrow nx(n-k) \neq y(n-k)$$

$\Rightarrow$  time variant

## Causality

A system is causal if the output at any time depends only on present and past values of the inputs and not on future values of the input.  $y(n) = x(-n)$  is non-causal because  $y(-1) = x(1)!$

## Stable versus Unstable Systems

A system is *Stable* if any bounded input produces bounded output (BIBO).

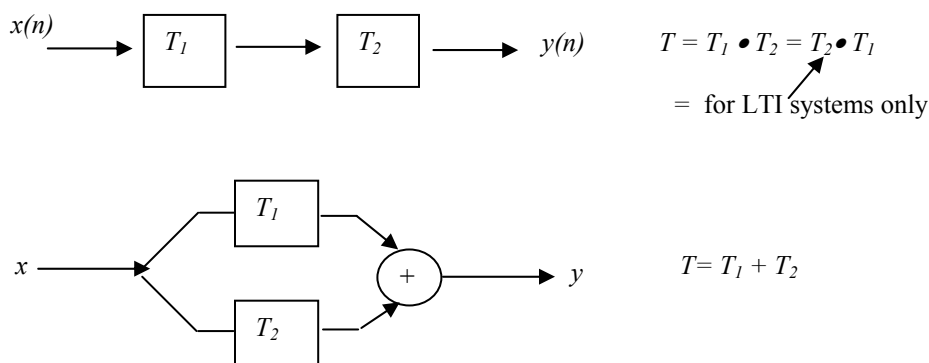
Otherwise, it is *unstable*.

## Linearity

A system is linear if

$$T(a_1x_1(n) + a_2x_2(n)) = a_1T[x_1(n)] + a_2T[x_2(n)]$$

two systems can be connected to each other in two ways:



## LTI Systems

LTI systems are the most important class of systems because the behavior of the system is known by knowing its response to unit sample input ( $x(n)$ ).

$$\begin{aligned}
 y(n) &= F\{x(n)\} = F\left[\sum_{K=-\infty}^{+\infty} x(K)\sigma(n-K)\right] \\
 \text{Then } &= \sum_{K=-\infty}^{+\infty} x(K)F\left[\underbrace{\sigma(n-K)}_{h(n-K)}\right] = \sum_{K=-\infty}^{+\infty} x(K)h(n-K) \\
 \Rightarrow y(n) &= x(n)*h(n) = h(n)*x(n) \\
 &= \sum_K x(K)h(n-K) = \sum_K h(K)x(n-K)
 \end{aligned}$$

## Causality of LTI System

$$\begin{aligned}
 y(n) &= \sum_{-\infty}^{\infty} h(k)x(n-k) = \sum_{-\infty}^{-1} h(k)x(n-k) + \sum_0^{\infty} h(k)x(n-k) \\
 &= \underbrace{[\dots + h(-2)x(n+2) + h(-1)x(n+1)]}_{\text{this depends on future values of } x(n)} + [h(0)x(n) + h(1)x(n-1) + \dots]
 \end{aligned}$$

this depends on future values of  $x(n)$

Hence, for a system to be casual, its  $h(n)$  must be zero for  $n < 0$ .

## Stability of LTI Systems

The BIBO system means:

$$\begin{aligned}
 |x(n)| &< M_x < \infty \rightarrow |y(n)|M_y < \infty \\
 y(n) &= \sum_{k=-\infty}^{+\infty} h(k)x(n-k) \\
 |y(n)| &= \left| \sum_{k=-\infty}^{+\infty} h(k)x(n-k) \right| \leq \sum_{k=-\infty}^{+\infty} |h(k)||x(n-k)| \leq \sum_k |h(k)|M_x
 \end{aligned}$$

Therefore if  $\sum_k |h(k)| < \infty$  meaning  $h(k)$  is absolutely summable, then  $|y(n)| < \infty \Rightarrow$  the system is stable. This condition is not only sufficient for stability of the system, but it is also necessary. Indeed we should show that if  $\sum_k |h(k)| = \infty$ , then a bounded input can produce an unbounded output. Let

$$x(n) = \begin{cases} \frac{h^*(-n)}{|h^*(-n)|} & h(n) \neq 0 \\ 0 & h(n) = 0 \end{cases} \quad \text{then } y(n) = \sum_k h(k)x(n-k) \text{ and}$$

$$y(0) = \sum_{-\infty}^{+\infty} h(k)x(-k) = \sum_k \frac{|h(k)|^2}{h(k)} = \sum |h(k)| = \infty \rightarrow y(0) = \infty \equiv \text{unbounded.}$$

The condition  $\sum_k |h(k)| < \infty$  also implies that  $h(n)$  goes to zero as  $n$  approaches  $\infty \Rightarrow$  the output of the system goes to zero as  $n \rightarrow \infty$  if the input is set to zero beyond  $n > n_o$ . In other words, a finite excitation to the system creates an output that eventually dies out (transient response).

Let's prove it. Suppose  $|x(n)| < M_x$  and also  $x(n) = 0$  for  $n > n_o$ . Then at any  $n = n_o + N$ , the system output is:

$$y(n_o + N) = \underbrace{\sum_{k=-\infty}^{N-1} h(k)x(n_o + N - k)}_{=0} + \sum_{k=N}^{\infty} h(k)x(n_o + N - k)$$

$\because x(n) = 0 \text{ for } n \geq n_o$

$$\text{Hence, } |y(n_o + N)| = \left| \sum_{k=N}^{\infty} h(k)x(n_o + N - k) \right| \leq \sum_{k=N}^{\infty} |h(k)||x(k)| \leq M_x \sum_{k=N}^{\infty} |h(k)|$$

Now as  $N \rightarrow \infty \lim_{N \rightarrow \infty} \sum_{n=N}^{\infty} |h(k)| = 0 \Rightarrow \lim_{N \rightarrow \infty} y(n_o + N) = 0$  Therefore, the transient response

goes to zero and hence the system is stable.

Example:  $h(n) = a^n u(n)$  determine the range that  $h(n)$  is stable.

1) System is causal for  $h(n) = 0$  for  $n < 0$

$$2) \sum_{k=-\infty}^{\infty} |h(k)| = \sum_{k=0}^{\infty} |a^k| = 1 + |a| + |a^2| + \dots$$

Now if  $|a| < 1$  this converges to  $\frac{1}{1-|a|}$  and the system would be stable but if  $|a| > 1$  the

system is unstable.

If  $h(n) = 0$  for  $n \geq M$  and  $n < 0$ , then the system is called FIR (Finite duration Impulse Response); otherwise the system is called IIR.

FIR system has a finite memory and it only looks at the input within a window.

### Analyzing the FIR and IIR System

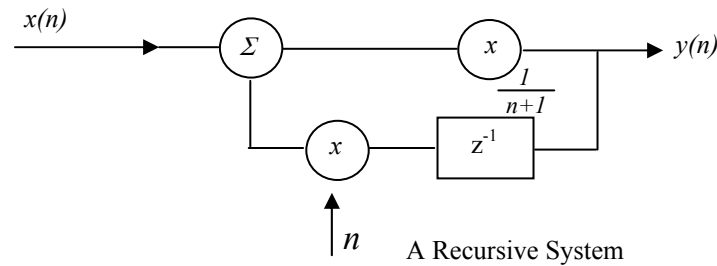
One approach is to use convolution sum but if the system is IIR, its practical implementation is impossible because it needs an infinite sum. Then to realize an IIR system, the solution is to use difference equations.

There is a sub-class of recursive and non-recursive systems. Consider a commulative average:

$$y(n) = \frac{1}{n+1} \sum_{k=0}^n x(k) \quad 0 \leq k \leq n$$

$$(n+1)y(n) = \sum_{k=0}^n x(k) = \sum_{k=0}^{n-1} x(k) + x(n) = ny(n-1) + x(n)$$

$$\Rightarrow y(n) = \frac{n}{n+1} y(n-1) + \frac{1}{n+1} x(n)$$



Is this system LTI? No, as it is time-variant because of multiplying by  $n$ .

### 2.4.2 LTI Systems Characterized by a Constant Coefficient Difference Equation

Consider  $y(n) = a y(n-1) + x(n)$

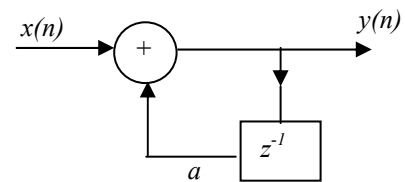
$$y(0) = ay(0) + x(1) = a^2 y(-1) + ax(0) + x(1)$$

↓

$$y(n) = a^{n+1} y(-1) + a^n x(0) + a^{n-1} x(1) + \dots$$

or

$$y(n) = \underbrace{a^{n+1} y(-1)}_{\text{zero-input response}} + \underbrace{\sum_{K=0}^n a^K x(n-K)}_{\text{zero-state or forced response}} = y_{zi}(n) + y_{zs}(n) **$$



The linearity applies to each of these responses separately. This system is linear and time-invariant.

*Impulse Response* –  $h(n)$  is simply equal to the zero-state response of the system.

$$y_{zs}(n) = \sum_{k=0}^n a^k x(n-k) \quad \text{Let } x(n) = \delta(n) \quad \text{then } \Rightarrow h(n) = \sum_{k=0}^n a^k \delta(n-k) = a^n \quad n \geq 0$$

Stability Example: is  $y(n) = a y(n-1) + x(n)$  stable?

Given a bounded input:  $|x(n)| \leq M_x < \infty$  for all  $n \geq 0$ , from (\*\*) we have

$$y(n) \leq |a^{n+1} y(-1)| + \left| \sum_{k=0}^n a^k x(n-k) \right| \leq |a^{n+1}| |y(-1)| + M_x \frac{1 - |a|^{n+1}}{1 - |a|} = M_y$$

if  $n$  is finite,  $M_y$  is finite but if  $n \rightarrow \infty$ ,  $M_y$  is bounded only if  $|a| < 1$ .